

# Theoretical and experimental demonstration of the transient responses of a circuit

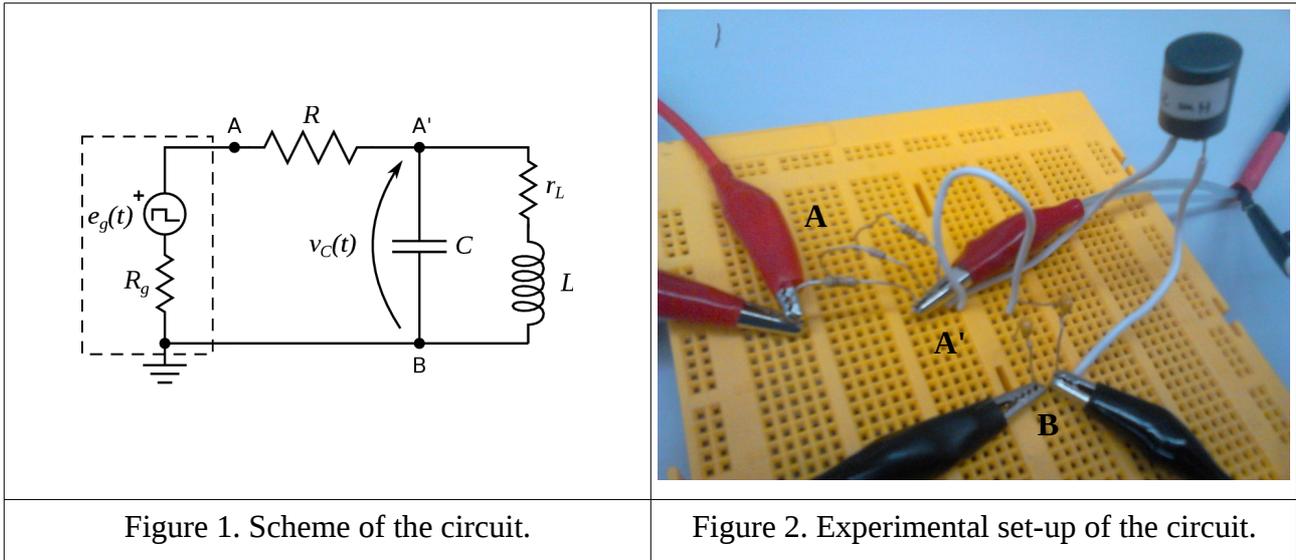
Practice proposed for Circuit Analysis, course 2016-17.

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## Problem

The scheme of the circuit shown in figure 1 corresponds to the real circuit shown in figure 2.



The circuit is connected to an external voltage source that applies the following “rec” function

$$e_g(t) = E_0 \left( u(t) - u\left(t - \frac{T}{2}\right) \right), \quad (1)$$

where  $T$  is the period of the signal generator which has a frequency of  $f$ , so that  $T=1/f$ , and the step height is  $E_0$ . The function generator has an internal resistance  $R_g=50\Omega$ . The coil of inductance  $L=22\text{mH}$ , has also an internal resistance  $r_L=25\Omega$ . For a given value of  $f$  and  $E_0$ , the values of the resistance,  $R$ , and the capacitance,  $C$ , are changed to see how they will affect the transient behavior of the voltage measured at the capacitor,  $v_C(t)$ .

The value of the capacitance,  $C$ , can be either  $470\text{nF}$  or  $10\text{nF}$ .

The value of the resistance,  $R$ , can be  $0$ ,  $100\Omega$ ,  $1000\Omega$  or  $10000\Omega$ .

For the theoretical solution it is to consider that time the voltage  $E_0$  is applied is much longer than the time the transient behavior happens ( $T/2 \gg \tau$ ).

## Theoretical solution

If  $T/2 \gg \tau$  then the initial conditions are close to zero:  $v_C(0)=0$ ,  $i_L(0)=0$ .

The circuit is then solved for  $t \geq 0$  and the resulting Laplace transform of the voltage at the

capacitor is:

$$V_C(s) = K \frac{\alpha + s}{s(s^2 + a_1s + a_0)} \left(1 - e^{-\frac{T}{2}s}\right), \quad (2)$$

where

$$K = \frac{E_0}{C(R + R_g)},$$

$$\alpha = \frac{r_L}{L},$$

$$a_1 = \left(\frac{r_L}{L} + \frac{1}{C(R + R_g)}\right),$$

$$a_0 = \left(1 + \frac{r_L}{R + R_g}\right) \frac{1}{LC}.$$

To obtain the inverse Laplace transform ( $\mathcal{L}^{-1}$ ) of  $V_C(s)$ , the following property can be used:

If  $V_C(s) = G(s)(1 - e^{-\frac{T}{2}s})$ , then

$$v_C(t) = g(t)u(t) - g\left(t - \frac{T}{2}\right)u\left(t - \frac{T}{2}\right), \quad (3)$$

being  $g(t) = \mathcal{L}^{-1}[G(s)]$ . Thus, it only remains to obtain the Laplace transform of

$$G(s) = K \frac{\alpha + s}{s(s^2 + a_1s + a_0)}. \quad (4)$$

From the characteristic polynomial that appears in eq.(4), the natural frequency,  $\omega_n$ , and the damping ration,  $\xi$ , can be obtained:

$$\omega_n = \sqrt{a_0}, \quad (5)$$

$$\xi = \frac{a_1}{2\sqrt{a_0}}. \quad (6)$$

Depending on the values of damping ration,  $\xi$ , or else, depending on the roots  $\{s_1, s_2\}$  of the characteristic polynomial  $s^2 + a_1s + a_0 = 0$ , the following different type of solutions are obtained:

- 1. Under-damped**, when  $0 < \xi < 1$ , and the roots are complex conjugated  $s_1 = -a + jb$ ,  $s_2 = -a - jb$ , being  $a = \xi\omega_n$ ,  $b = \omega_n\sqrt{1 - \xi^2}$ . The following inverse Laplace transform (Nr. 28 of the table) is obtained:

$$g(t) = A + Be^{-at} \sin(bt + \phi), \quad (7)$$

where

$$A = \frac{K\alpha}{a^2 + b^2} = g(t \rightarrow \infty), \text{ which is the voltage at } r_L.$$

$$B = \frac{K}{b} \sqrt{\frac{(\alpha - a)^2 + b^2}{a^2 + b^2}},$$

$$\phi = \arctan \frac{b}{\alpha - a} - \arctan \frac{b}{-a}.$$

How fast the under-damped transient behavior decreases is given by the inverse of the time constant or settling time,  $\tau$ , obtained from the real part of the roots of the characteristic polynomial. For the under-damped systems:

$$\tau = \frac{1}{a} = \frac{1}{\xi\omega_n}, \quad (8)$$

This settling time is always larger than the one obtained for the over-damped and critically-damped solutions. And the oscillation period is given by the imaginary part of the roots:

$$T_u = \frac{2\pi}{b} = \frac{2\pi}{\omega_n\sqrt{\xi^2 - 1}}, \quad (9)$$

2. **Over-damped**, when  $\xi > 1$ , and the roots  $s_1 = -a, s_2 = -b$  are real, negative and different, being  $a = \xi\omega_n - \omega_n\sqrt{\xi^2 - 1}$ ,  $b = \xi\omega_n + \omega_n\sqrt{\xi^2 - 1}$ . The following inverse Laplace transform (Nr. 11 of the table) is obtained:

$$g(t) = A + Be^{-at} + De^{-bt}, \quad (10)$$

where

$$A = \frac{K\alpha}{ab},$$

$$B = -\frac{Kb(\alpha - a)}{a(b - a)},$$

$$D = \frac{K(\alpha - b)}{b(b - a)}.$$

It is easy to check that  $g(t = 0) = A + B + D = v_C(0) = 0$ .

3. **Critically damped**, when  $\xi = 1$ , and the roots are real and equal,  $s_1 = s_2 = -a = -\xi\omega_n$ . The obtained inverse Laplace transform (Nr. 34) is:

$$g(t) = A + (B + Dt)e^{-at}. \quad (11)$$

where

$$A = \frac{K\alpha}{a^2},$$

$$B = -A \Rightarrow g(0) = 0,$$

$$D = \frac{K(a - \alpha)}{a}.$$

The value of  $R$  for which the solution is critically damped (when  $a_1 = 2\sqrt{a_0}$ ) is

$$R_c = 2 \left[ 2\alpha C + \sqrt{(2\alpha C)^2 - 4\frac{C}{L}(\alpha C - 4)} \right]^{-1} - R_g. \quad (12)$$

## Experimental set-up

The function generator is shown in figure 3, and is connected to the circuit shown in figure 2 at the

points A and B (B is GND), and the channel 1 of the oscilloscope (shown in figure 4) is connected to the points A' and B. To avoid flickering and replications of the visualized signal, it is necessary to introduce a trigger signal into the oscilloscope provided by the same signal of the function generator. The input signal of the function generator can also be visualized by connecting the channel 2 of the oscilloscope to the points A and B.



Figure 3. Function generator.

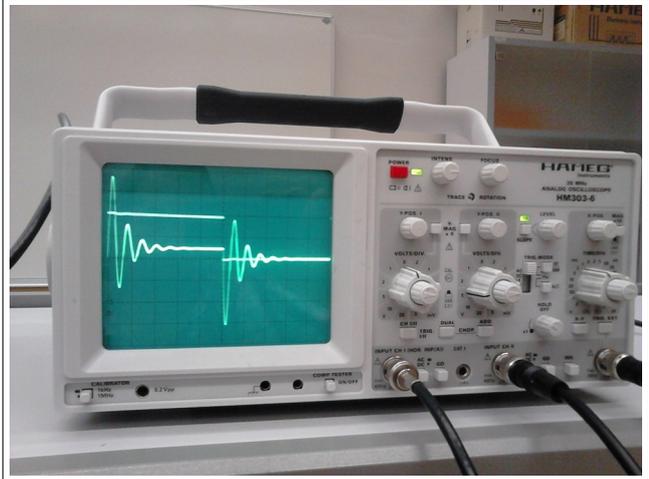


Figure 4. Oscilloscope.

### Comparison between theoretical and experimental results

The theoretical representations of eq.(3) (with their corresponding type of solutions for  $g(t)$  given in eq.(7), (10) or (11)) are obtained with MATLAB so that the time evolution of  $v_C(t)$  is represented as if it were displayed by the screen of the oscilloscope, according to the VOLTS/DIV. and TIME/DIV. adjustments, but the time interval is restricted between 0 and  $T=1/f$ .

Generate a “rec” signal (eq. (1)) using the function generator with, for example,  $E_0 = 10V$  and  $f=100$  Hz. (In the real experiment adjust the voltage with channel 1 of the oscilloscope before connecting the function generator to the circuit).

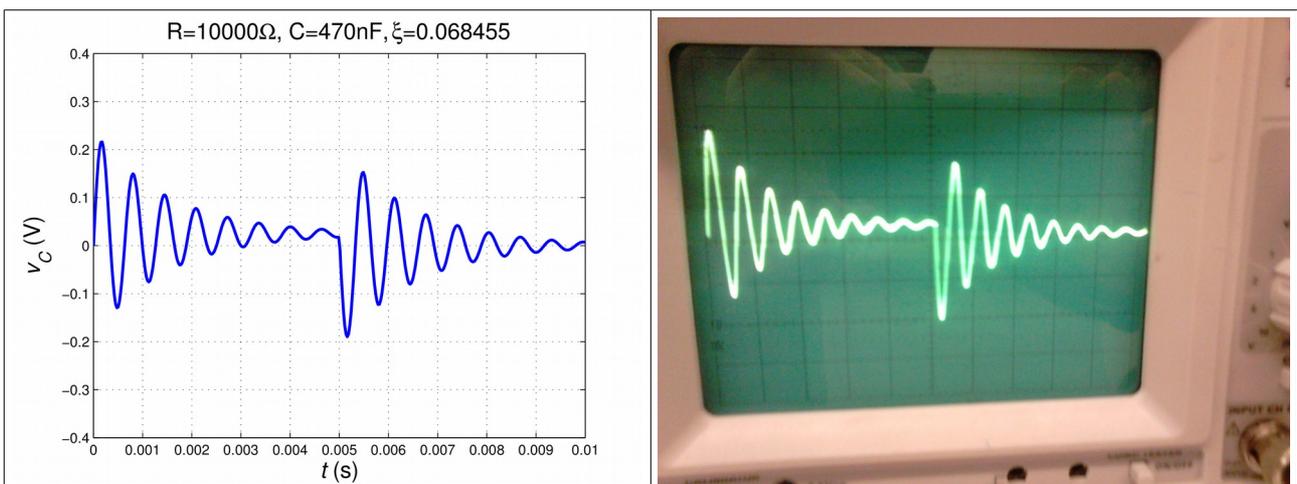


Figure 5(a). VOLT/DIV=0.1V, TIME/DIV=1ms, f=100Hz.

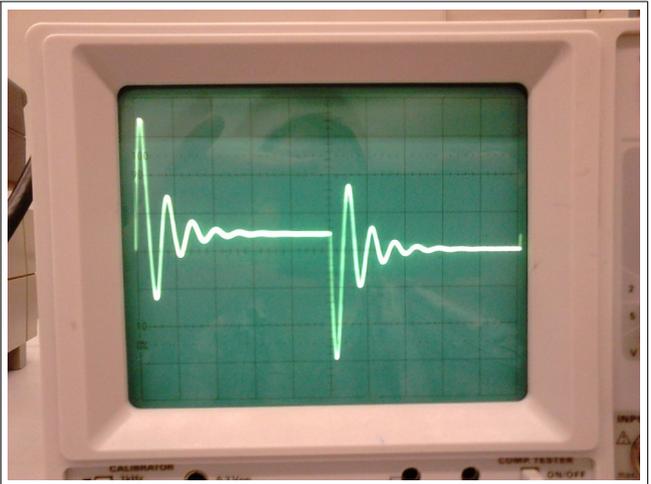
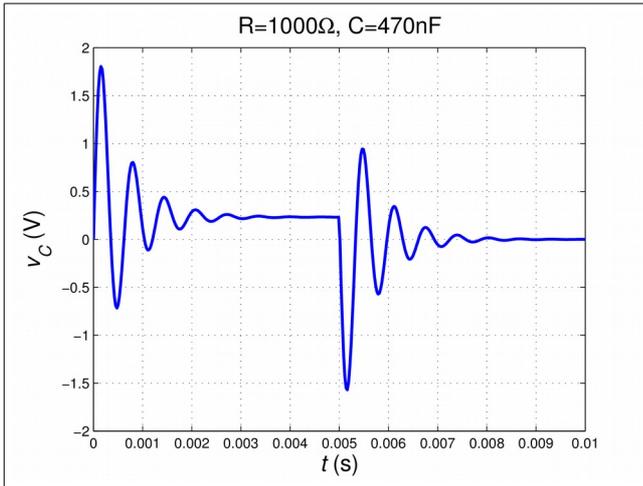


Figura 5(b). VOLT/DIV=0.5V, TIME/DIV=1ms, f=100Hz.

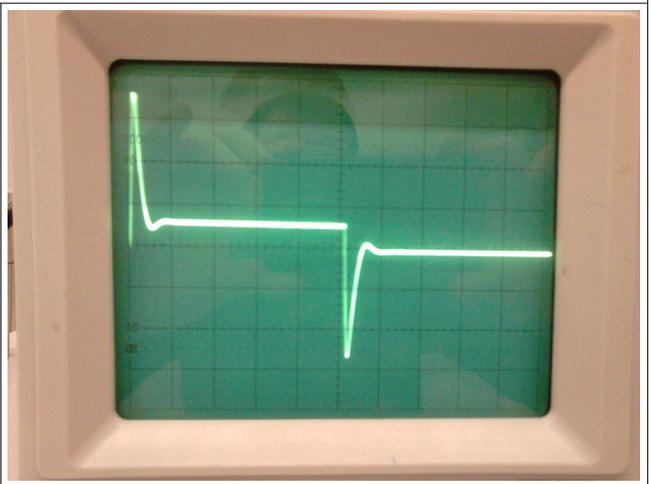
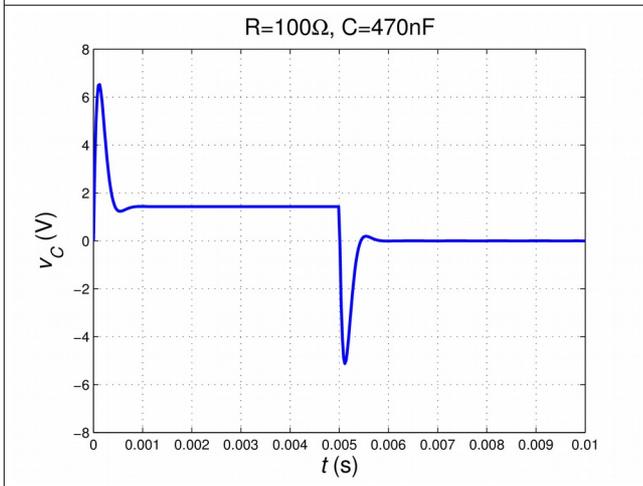


Figura 5(c). VOLT/DIV=2V, TIME/DIV=1ms, f=100Hz.

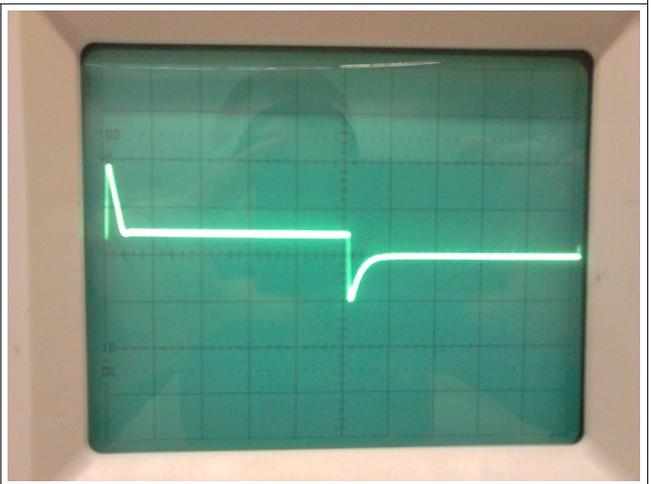
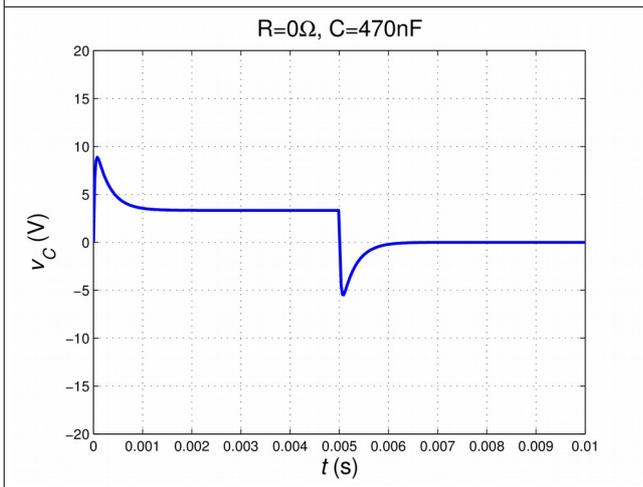


Figura 5(d). VOLT/DIV=5V, TIME/DIV=1ms, f=100Hz.

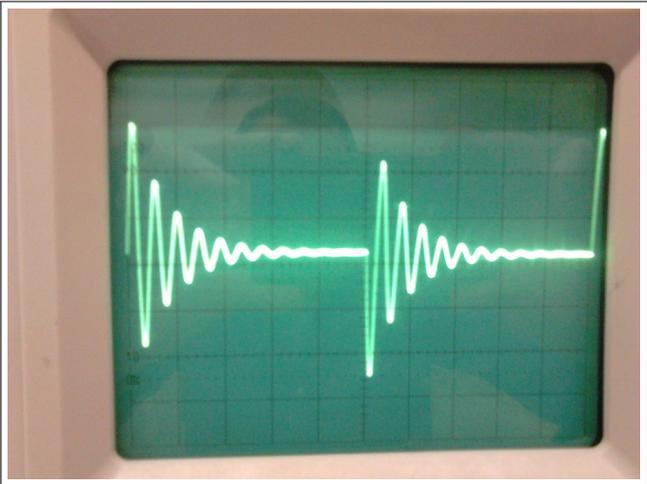
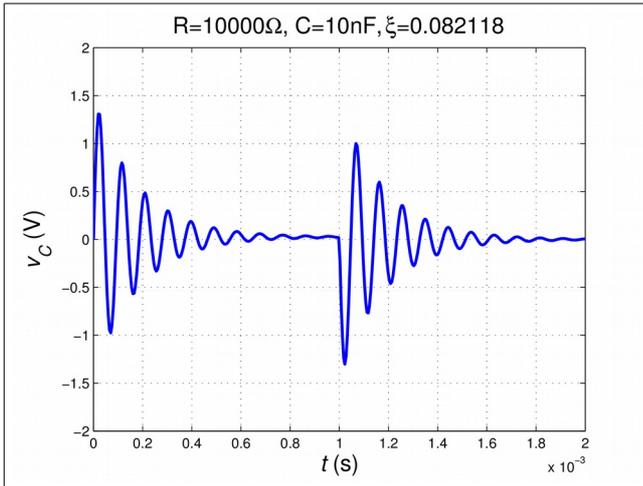


Figura 5(e). VOLT/DIV=0.5V, TIME/DIV=0.2ms,  $f=500\text{Hz}$ .

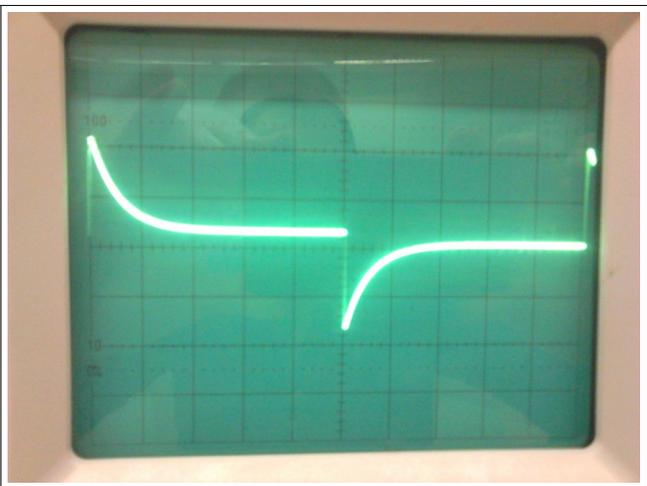
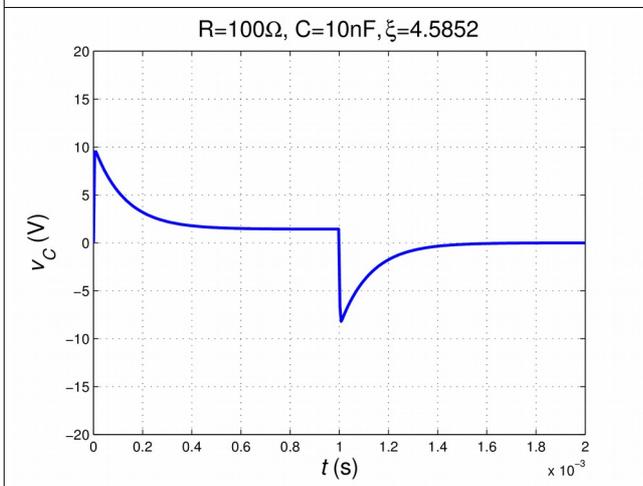


Figura 5(f). VOLT/DIV=5V, TIME/DIV=0.2ms,  $f=500\text{Hz}$ .

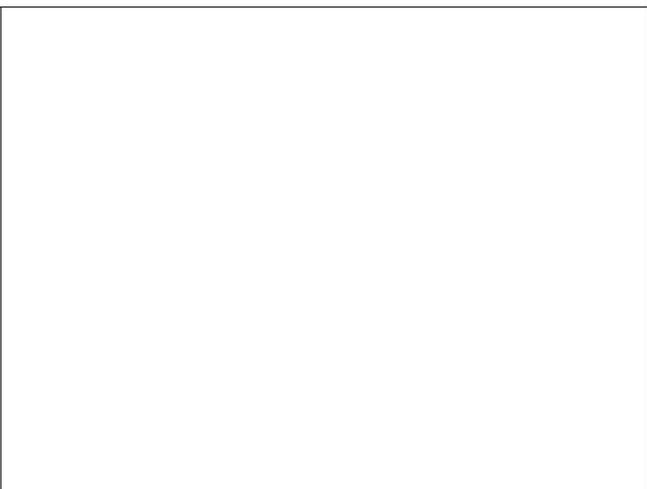
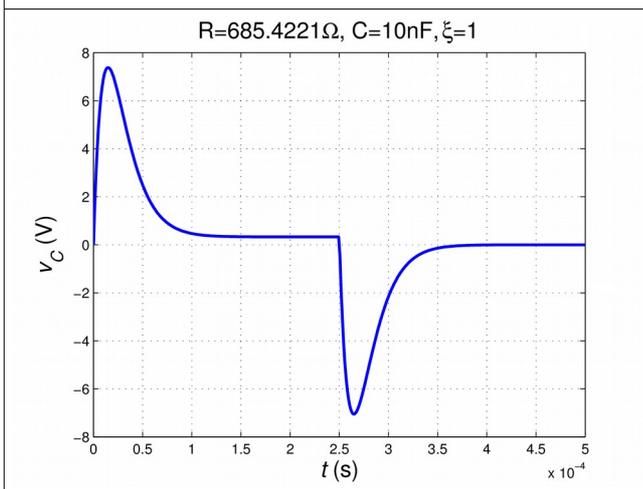


Figura 5(g). VOLT/DIV=2V, TIME/DIV=50μs,  $f=2000\text{Hz}$ .

## Matlab program

The following program for MATLAB® can be used for graphically representing the theoretical obtained solutions. (Copy and paste in your .m file).

```
%% Transient_response_Experiment
% Visualisation of the transient responses of a rectangular pulse.
% The theoretical results of the corresponding experiment are computed here
% to demonstrate that the theory can be used to explain and predict the real
% phenomena.
%% Experimental Set-up:
% Function generator {eg(t)+Rg} connected to {R + C//{rL+L}}
% Oscilloscope (e.g. channel 1) connected to C: vC(t) is displayed
% (Ground connection have to be connected to the same ground of the source)
% The oscilloscope has to be triggered with the output signal of the
% function generator.
% Channel 2 of the oscilloscope can be connected to the source output to
% also visualize the applied external "Rec" signal

%% DATA
%Circuits components
L=22*10^(-3); %H
rL=25; %Ohm, internal resistance of L: {rL+L}
%C=470*10^(-9);
C=10*10^(-9);
R=10000; %Ohm
%R=1000;
%R=100;
%R=685.4221
%R=0
%Function generator: "Rec" pulse: eg(t)=E0*(u(t)-u(t-T/2)), T=1/f
E0=10; %V
f=500; %Hz
Rg=50; %Ohm, internal resistance of the function generator {E+Rg}
%Oscilloscope set-up:
Tdiv=0.2*10^(-3); %s
Vdiv=0.5; %V

%% SOLUTION
%Voltage at C: vC(t)

%Time interval for vC(t) is one period of the function generator
T=1/f;
t = [0:T/399:T];

%The Laplace transform of vC(t):
%VC(s)=K*(s+alpha)*(1-exp(-T*s/2))/[s*(s^2+a1*s+a0)],
%being:
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```

K = E0./(C*(R+Rg));
alpha = rL./L;
a1 = rL/L + 1./(C.*(R+Rg));
a0 = (1+rL/(R+Rg))./(L*C);

wn = sqrt(a0); %Natural frequency
D = a1./(2*wn) %Damping ratio
tau = 1/(D*wn) %Settling time

%Value of R for Critically damped solution (when a1=2*sqrt(a0))
a = -2*rL*C/L; b = (rL*C/L)^2-4*C/L;
Rc(1) = 2/(-a + sqrt(a^2-4*b))-Rg; Rc(2) = 2/(-a - sqrt(a^2-4*b))-Rg;
Rc = Rc(Rc>0) %it is the positive solution
if isempty(Rc)==1
    display('there is no R to get critically damped solution')
end
clear a b

%The inverse Laplace transform L-1 { }
%G(s)=K*(s+alpha)/[s*(s^2+a1*s+a0)] -> g(t)=L-1{G(s)}
%VC(s)=G(s)*(1-exp(-T*s/2)) -> vC(t)=g(t)u(t)-g(t-T/2)u(t-T/2);
u=ones(size(t)); %u=u(t)
uT=u; uT(1:round(length(t)/2))=0; %uT=u(t-T)
if D<1; %s1 and s2 are complex conjugated (Underdamped)
    a = D.*wn;
    b = wn*sqrt(1-D^2); %Damped natural frequency
    %s1=-a+i*b; s2=-a-i*b;
    phi = atan2(b,alpha-a)-atan2(b,-a);
    A = K*alpha/(a^2+b^2);
    B = (K/b).*sqrt(((alpha-a)^2+b^2)/(a^2+b^2));
    vC1 = (A + B.*exp(-a.*t).*sin(b.*t+phi)).*u; %table Nr. 28
    vC2 = - (A + B.*exp(-a.*(t-T/2)).*sin(b.*(t-T/2)+phi)).*uT;
    vC2(isnan(vC2)==1)=0;
    vC=vC1+vC2;
elseif D>1 %s1 and s2 are real, negative and different (Overdamped)
    a = D*wn-wn.*sqrt(D^2-1); %a=-s1
    b = D*wn+wn.*sqrt(D^2-1); %b=-s2
    A = b.*(alpha-a)./(b-a);
    B = a.*(alpha-b)./(b-a);
    vC1 = (K/(a*b)).*(alpha - A.*exp(-a.*t) + B.*exp(-b.*t)).*u; %table Nr. 11
    vC2 = -(K/(a*b)).*(alpha - A.*exp(-a.*(t-T/2)) + ...
        B.*exp(-b.*(t-T/2))).*uT;
    vC2(isnan(vC2)==1)=0;
    vC=vC1+vC2;
elseif D==1 %s1=s2 are real, negative and equal (Critically damped)
    a = D.*wn; %s1=-a
    vC1 = (K/a^2).*(alpha - alpha.*exp(-a.*t) + ...
        a*(a-alpha).*t.*exp(-a.*t)).*u; %table Nr. 34
    vC2 = (K/a^2).*(alpha - alpha.*exp(-a.*(t-T/2)) + ...
        a*(a-alpha).*(t-T/2).*exp(-a.*(t-T/2))).*uT;

```

```

vC2(isnan(vC2)==1)=0;
vC=vC1+vC2;
end
%(The undamped solution can not be obtained with the experiment)

%% Representation (con ajustes del osciloscopio)
figure
plot(t,vC,'LineWidth',2)
axis([0 10*Tdiv -4*Vdiv 4*Vdiv])
grid on
xlabel('\itt (s)','FontSize',16)
ylabel('\itv_C (V)','FontSize',16)
title(['R=',num2str(R),'\Omega, C=',num2str(C*10^9),...
'nF, \xi=',num2str(D),', \tau=',num2str(tau),'s'],'FontSize',16)

```

## Questions

1. Deduce equation (2).

Consider the following cases:

a)  $R=10\text{k}\Omega$ ,  $C=470\text{nF}$ .

b)  $R=1\text{k}\Omega$ ,  $C=470\text{nF}$ .

c)  $R=100\Omega$ ,  $C=470\text{nF}$ .

d)  $R=0$ ,  $C=479\text{nF}$

e)  $R=10\text{k}\Omega$ ,  $C=10\text{nF}$ .

f)  $R=100\Omega$ ,  $C=10\text{nF}$ .

g)  $R=685.42\Omega$ ,  $C=10\text{nF}$ .

2. Obtain the temporal expression of  $v_C(t)$  for the cases e), f), and g).

3. Perform the experiments of the measurements of  $v_C(t)$  for the cases a) to f), take a picture from the screen of the oscilloscope. Compare them with the theoretically obtained results and answer the following questions:

3.1 For the same value of  $C$ , does the damping ratio ( $\xi$ ) increase or decrease for smaller  $R$ ? How can this be explained?.

3.2 How affects the value of  $C$  to the oscillation for the under-damped cases.

3.3 Way decreases the transient response of  $v_C(t)$  for  $0 < t < T/2$  not to zero?. (or, the voltage at  $C$  tends to the voltage at  $\zeta \dots$ ?)

7. Considering  $C=470\text{nF}$ , what is the value of  $R$  that produces the critically damped behavior.

8. What happens when the frequency of the function generator increases?. (or, what happens when  $T/2$  becomes closer to  $\tau$ ?)